

## Section 10.2 Sinusoidal Sources

### P10.2-1

Given:

$$v_1(t) = 5 \cos(150t + 30^\circ) \text{ V}$$

$$v_2(t) = 4 \cos(150t - 60^\circ) \text{ V}$$

So,

$$\theta_1 = 30^\circ$$

$$\theta_2 = -60^\circ$$

The period of the sinusoids is given by,

$$150 = \frac{2\pi}{T}$$

$$T = 42 \text{ ms}$$

Compare  $v_2(t)$  to  $v_1(t)$

$$\theta_2 - \theta_1 = -60 - 30 = -90^\circ = -\frac{\pi}{2} \text{ rad}$$

The minus sign (–) indicates a delay rather than an advance,  
Convert the angle to time.

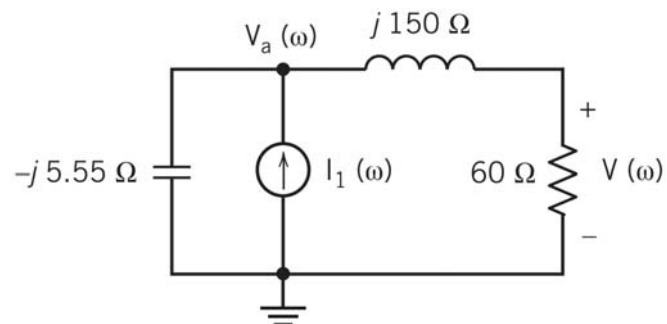
$$\theta_2 - \theta_1 = 2\pi \frac{t_d}{T}$$

$$\begin{aligned} t_d &= \frac{(\theta_2 - \theta_1)T}{2\pi} \\ &= \frac{(-\pi/2)(0.042)}{2\pi} \\ &= -0.0105 \text{ s} \end{aligned}$$

### P10.4-5

**Solution:**

The figure below shows the given circuit:



The impedance of inductor is:

$$\mathbf{Z}_L = j\omega L$$

Substitute 6000 rad/s for  $\omega$ , and 25 mH for  $L$ ,

$$\begin{aligned}\mathbf{Z}_L &= j(6000 \text{ rad/s})(25 \text{ mH}) \left( \frac{10^{-3} \text{ H}}{1 \text{ mH}} \right) \\ &= j150 \Omega\end{aligned}$$

The impedance of the capacitor is:

$$\mathbf{Z}_C = \frac{-j}{\omega C}$$

Substitute 6000 rad/s for  $\omega$ , and 30  $\mu\text{F}$  for  $C$ ,

$$\begin{aligned}\mathbf{Z}_C &= \frac{-j}{(6000 \text{ rad/s})(30 \mu\text{F}) \left( \frac{10^{-6} \text{ F}}{1 \mu\text{F}} \right)} \\ &= -j5.55 \Omega\end{aligned}$$

The phasor representation of current  $i(t) = 180 \cos(6000t)$  is:

$$\begin{aligned}\mathbf{I}(\omega) &= 180 \angle 0^\circ \\ &= 180 \cos(0^\circ) - j180 \sin(0^\circ) \\ &= 180 \text{ mA}\end{aligned}$$

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Apply Nodal analysis at node 'a':

$$\begin{aligned}\frac{\mathbf{V}_a(\omega)}{\mathbf{Z}_C} + \frac{\mathbf{V}_a(\omega)}{\mathbf{Z}_L + R} &= \mathbf{I}(\omega) \\ \mathbf{V}_a(\omega) &= \frac{\mathbf{I}(\omega)}{\left( \frac{1}{\mathbf{Z}_C} + \frac{1}{\mathbf{Z}_L + R} \right)} \\ &= \frac{\mathbf{I}(\omega)(\mathbf{Z}_C)(\mathbf{Z}_L + R)}{\mathbf{Z}_C + \mathbf{Z}_L + R}\end{aligned}$$

Substitute the values,

$$\begin{aligned} \mathbf{V}_a(\omega) &= \frac{(180 \text{ mA})(-j5.55 \Omega)(j150 \Omega + 60 \Omega)}{-j5.55 \Omega + j150 \Omega + 60 \Omega} \\ &= \frac{150 - j60 \text{ A}\Omega^2}{60 \Omega + j144.5 \Omega} \end{aligned}$$

The voltage division rule between the inductor and the resistor gives the voltage  $\mathbf{V}(\omega)$  as:

$$\mathbf{V}(\omega) = \left( \frac{R}{R + \mathbf{Z}_L} \right) \mathbf{V}_a(\omega)$$

Substitute the values,

$$\begin{aligned} \mathbf{V}(\omega) &= \left( \frac{60 \Omega}{60 \Omega + j150 \Omega} \right) \left( \frac{150 - j60 \text{ A}\Omega^2}{60 \Omega + j144.5 \Omega} \right) \\ &= \frac{9000 - j3600 \text{ A}\Omega^3}{-18075 + j17670 \Omega^2} \\ &= \frac{1800 - j720 \text{ A}\Omega^3}{-3615 + j3534 \Omega^2} \\ &= -0.3542 - j0.1471 \text{ V} \end{aligned}$$

The phase angle of the voltage  $\mathbf{V}(\omega)$  is:

$$\begin{aligned} \theta &= \tan^{-1} \left( \frac{-0.1471}{-0.3542} \right) \\ &= -157.45^\circ \end{aligned}$$

The amplitude of voltage  $\mathbf{V}(\omega)$  is:

$$\begin{aligned} V &= \sqrt{(0.3542 \text{ V})^2 + (0.1471 \text{ V})^2} \\ &= 0.3835 \text{ V} \end{aligned}$$

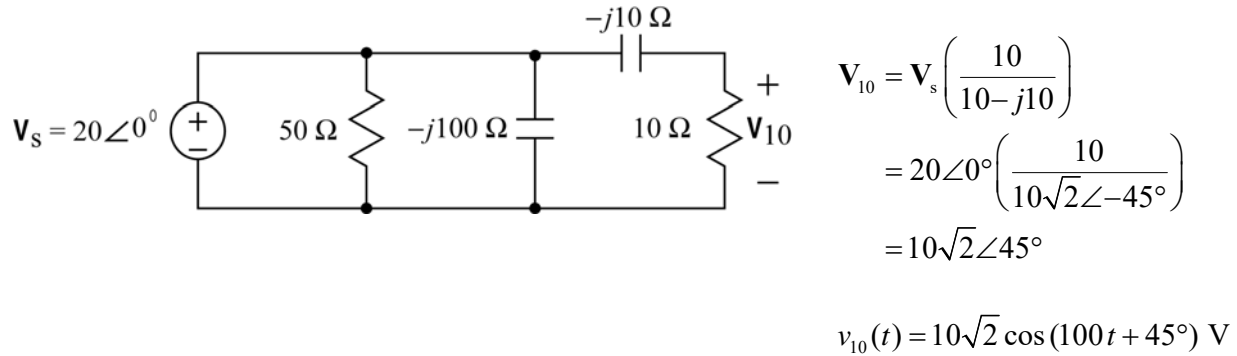
The voltage across the resistor is:

$$\begin{aligned} v(t) &= V \cos(6000t + \theta) \\ &= 0.3835 \cos(6000t - 157.45^\circ) \end{aligned}$$

Therefore, the voltage across the resistor is  $0.3835 \cos(6000t - 157.45^\circ)$ .

**P 10.5-12**

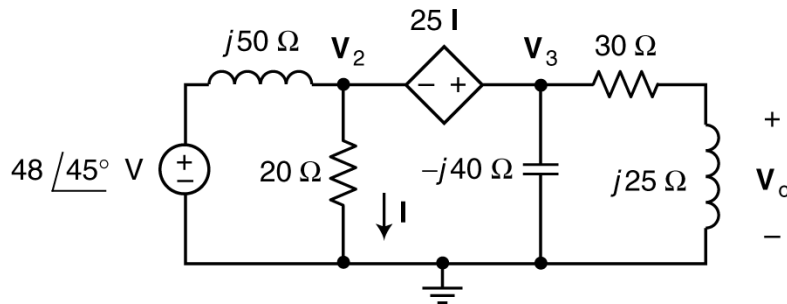
**Solution:**



**P10.6-1**

**Solution:**

Represent the circuit in the frequency domain as



The node voltages are  $48 \angle 45^\circ = \mathbf{V}_1$ ,  $\mathbf{V}_2$ ,  $\mathbf{V}_3$  and  $\mathbf{V}_o$ . Express the dependent source voltage in terms of the node voltages:

$$\mathbf{V}_3 - \mathbf{V}_2 = 25 \mathbf{I} = 25 \left( \frac{\mathbf{V}_2}{20} \right) \Rightarrow \mathbf{V}_3 = 2.25 \mathbf{V}_2$$

Apply KCL to the supernode corresponding to the CCVS to get

$$\begin{aligned} \frac{48 \angle 45^\circ - \mathbf{V}_2}{j50} &= \frac{\mathbf{V}_2}{20} + \frac{\mathbf{V}_3 - \mathbf{V}_o}{30} + \frac{\mathbf{V}_3}{-j40} \\ \frac{48 \angle 45^\circ}{j50} &= \frac{\mathbf{V}_2}{j50} + \frac{\mathbf{V}_2}{20} + \frac{\mathbf{V}_3 - \mathbf{V}_o}{30} + \frac{\mathbf{V}_3}{-j40} \\ \frac{48 \angle 45^\circ}{j50} &= \left( \frac{1}{j50} + \frac{1}{20} \right) \mathbf{V}_2 + \left( \frac{1}{30} + \frac{1}{-j40} \right) \mathbf{V}_3 - \frac{1}{30} \mathbf{V}_o \end{aligned}$$

$$\frac{48\angle 45^\circ}{j50} = \left( \frac{1}{j50} + \frac{1}{20} \right) \mathbf{V}_2 + \left( \frac{1}{30} + \frac{1}{-j40} \right) 2.25\mathbf{V}_2 - \frac{1}{30} \mathbf{V}_o$$

$$\frac{48\angle 45^\circ}{j50} = \left( \frac{1}{j50} + \frac{1}{20} + \frac{2.25}{30} + \frac{2.25}{-j40} \right) \mathbf{V}_2 - \frac{1}{30} \mathbf{V}_o$$

Apply KCL at the right node of the 30  $\Omega$  resistor to get

$$\frac{\mathbf{V}_3 - \mathbf{V}_o}{30} = \frac{\mathbf{V}_o}{j25} \Rightarrow 0 = \left( -\frac{1}{30} \right) 2.25\mathbf{V}_2 + \left( \frac{1}{30} + \frac{1}{j25} \right) \mathbf{V}_o$$

In matrix form

$$\begin{bmatrix} \frac{1}{j50} + \frac{1}{20} + \frac{2.25}{30} + \frac{2.25}{-j40} & -\frac{1}{30} \\ -\frac{2.25}{30} & \frac{1}{30} + \frac{1}{j25} \end{bmatrix} \begin{bmatrix} \mathbf{V}_2 \\ \mathbf{V}_o \end{bmatrix} = \begin{bmatrix} \frac{48\angle 45^\circ}{j50} \\ 0 \end{bmatrix}$$

Solving, perhaps using MATLAB,

$$\begin{bmatrix} \mathbf{V}_2 \\ \mathbf{V}_o \end{bmatrix} = \begin{bmatrix} 10.18\angle -44.6^\circ \\ 14.67\angle 5.6^\circ \end{bmatrix} \text{ V}$$

## MATLAB

```
>> A=[1/(j*50)+1/20+2.25/30+2.25/(-j*40) -1/30;-2.25/30 1/30+1/(j*25)]
```

```
A =
```

```
0.1250 + 0.0363i -0.0333 + 0.0000i
```

```
-0.0750 + 0.0000i 0.0333 - 0.0400i
```

```
>> V=inv(A)*[48*exp(j*45*pi/180)/(j*50);0]
```

```
V =
```

```
7.2496 - 7.1526i
```

```
14.5998 + 1.4265i
```

```
>> abs(V)
```

```
ans =
```

```
10.1841
```

```
14.6694
```

```
>> angle(V)*180/pi
```

```
ans =
```

```
-44.6139
```

```
5.5805
```

**P10.7-3**

**Solution:**

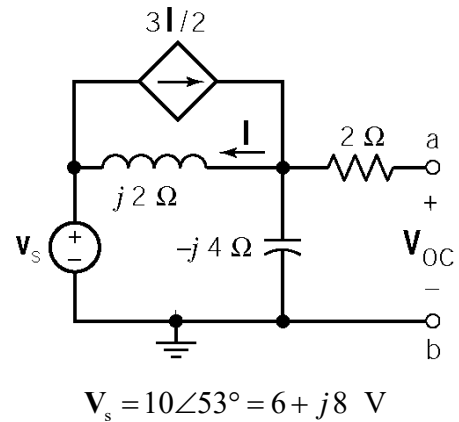
First, determine  $V_{oc}$ :

The node equation is:

$$\frac{V_{oc}}{-j4} + \frac{V_{oc} - (6 + j8)}{j2} - \frac{3}{2} \left( \frac{V_{oc} - (6 + j8)}{j2} \right) = 0$$

$$\left( -\frac{1}{j4} + \frac{1}{j2} - \frac{3}{j4} \right) V_{oc} = \frac{6 + j8}{j2} - \frac{3}{2} \left( \frac{6 + j8}{j2} \right)$$

$$V_{oc} = 3 + j4 = 5 \angle 53.1^\circ \text{ V}$$



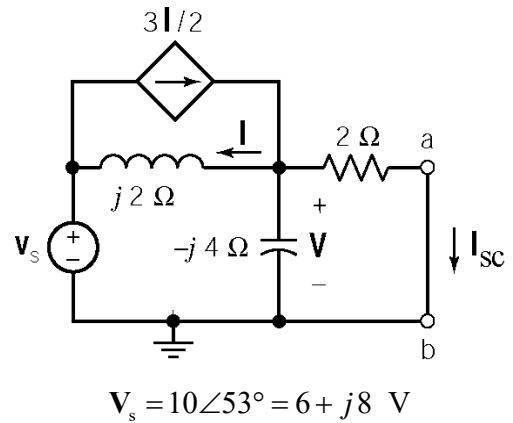
Next, determine  $I_{sc}$ :

The node equation is:

$$\frac{V}{2} + \frac{V}{-j4} + \frac{V - (6 + j8)}{j2} - \frac{3}{2} \left[ \frac{V - (6 + j8)}{j2} \right] = 0$$

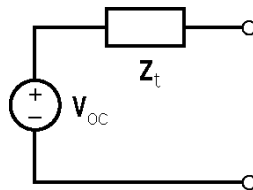
$$V = \frac{3 + j4}{1 - j}$$

$$I_{sc} = \frac{V}{2} = \frac{3 + j4}{2 - j2}$$



The Thevenin impedance is  $Z_T = \frac{V_{oc}}{I_{sc}} = (3 + j4) \left( \frac{2 - j2}{3 + j4} \right) = 2 - j2 \ \Omega$

The Thevenin equivalent is

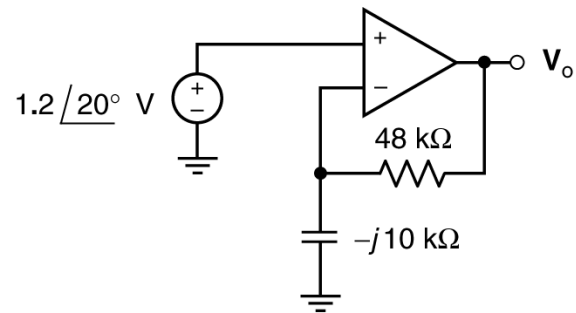


(checked: LNAP 7/18/04)

**P10.10-1**

**Solution:**

Represent the circuit in the frequency domain as



Recognizing this circuit as a noninverting amplifier, we can write

$$\mathbf{V}_o = \left( 1 + \frac{48}{-j10} \right) (1.2 \angle 20^\circ) = (1 + j4.8) (1.2 \angle 20^\circ) = 5.88 \angle 98^\circ \text{ V}$$

In the time domain  $v_o(t) = 5.88 \cos(400t + 98^\circ) \text{ V}$

(Checked using LNAPAC 3/15/12)