

MATLAB Problem

EFS.m

```
function [C0, Cn] = EFS(N, T)
%EFS Exponential Fourier Series
% returns the coefficients of the exponential Fourier
% Series of a periodic function described in the
% MATLAB function named my_periodic_function.m
%
% N = the number of harmonic frequencies
% T = the period of the periodic function
%
% C0=average value
% Cn(1)=C1, Cn(2)=C2, ..., Cn(N)=CN

% -----
% Obtain a list of equally spaced instants of time
% -----
n=2*N;
t=linspace(0,T,n+1);
t(end)=[];
% -----
% Obtain values of f(t) at those instants of time
% -----
f=my_squarewave(t,T);
```

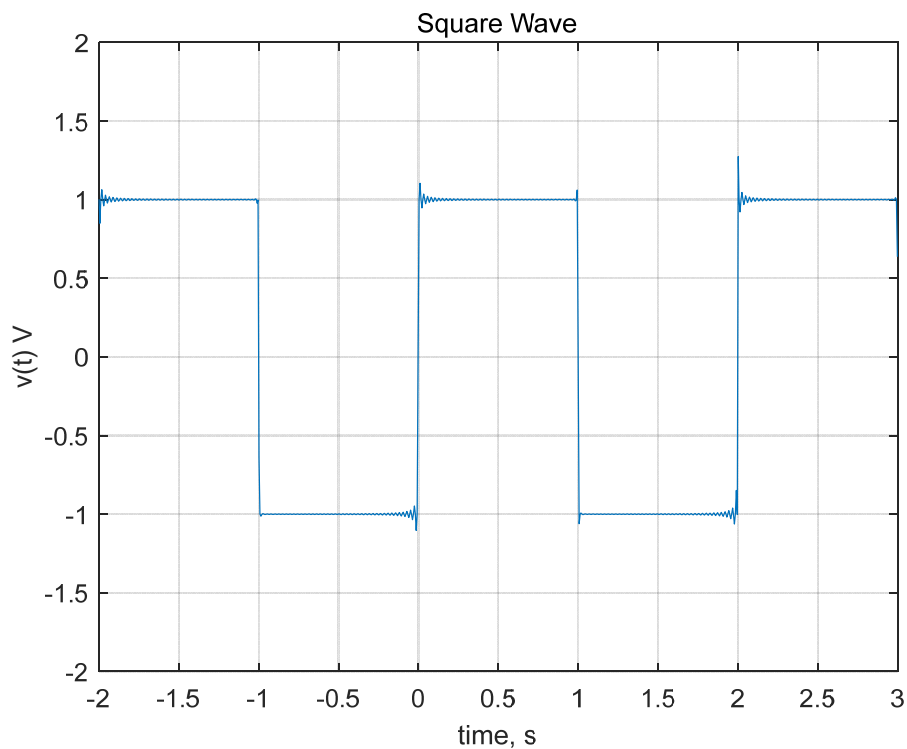
my_squarewave.m

```
function f = my_squarewave(t, T)
% % sawtooth with amplitude A and period T
% A=8;
% f=A*rem(t,T)/T;
% f(f==0 |f==A) = A/2;

% squarewave
A=1;
for k=1:length(t)
    if (t(k)<T/2) f(k)=A;
    elseif (t(k)>T/2) f(k)=-A;
    else f(k)= 0;
    end
end
```

testEFS.m

```
% testEFS.m
% -----
% Obtain a list of equally spaced instants of time
% -----
T=2;      % period
w0=2*pi/T; % fundamental frequency, rad/s
t0=-T;    % initial time
tf=1.5*T; % final time
dt=tf/512; % time increment
t=-T:dt:tf; % time, s
% -----
% Call EFS to get exponential Fourier coefficients
% -----
N=256; %Number of harmonic frequencies
[C0, Cn] = EFS(N,T);
% -----
% Approximate the function by its Fourier series
% -----
v = C0*ones(size(t)); % initialize v(t) as vector
for n=1:N
    v = v + Cn(n)*exp(j*n*w0*t) + Cn(n)'+exp(-j*n*w0*t);
end
% -----
% Plot the Fourier series |
% -----
plot(t, v)
axis([t0 tf -2 2])
```



P 15.5-2**Solution:**

$$\begin{aligned} C_n &= \frac{1}{2T} \int_0^{2T} \left(\frac{A}{2T} t \right) e^{-j \frac{2\pi}{T} nt} dt \\ &= \frac{A}{T^2} \int_0^{2T} t e^{-j(\pi/T)nt} dt \end{aligned}$$

Recall the formula for integrating by parts: $\int_{t_1}^{t_2} u dv = u v \Big|_{t_1}^{t_2} - \int_{t_1}^{t_2} v du$. Take $u = t$ and

$dv = e^{-j \frac{\pi}{T} nt} dt$. When $n \neq 0$, we get

$$\begin{aligned} C_n &= \frac{A}{4T^2} \left(\frac{t e^{-j(\pi/T)nt}}{-j(\pi/T)n} \Big|_0^{2T} + \frac{1}{j(\pi/T)n} \int_0^{2T} e^{-j(\pi/T)nt} dt \right) \\ &= \frac{A}{4T^2} \left(\frac{t e^{-j(\pi/T)nt}}{-j(\pi/T)n} + \frac{e^{-j(\pi/T)nt}}{(\pi/T)^2 n^2} \Big|_0^{2T} \right) \\ &= \frac{A}{4T^2} \left(\frac{2T e^{-j2\pi n}}{-j2\pi n} + \frac{e^{-j2\pi n} - 1}{(\pi/T)^2 n^2} \right) \\ &= j \frac{A}{2\pi n} \end{aligned}$$

Now for $n = 0$ we have

$$C_0 = \frac{1}{2T} \int_0^{2T} \frac{A}{2T} t dt = \frac{A}{2}$$

Finally,

$$f(t) = \frac{A}{2} + j \frac{A}{2\pi} \sum_{\substack{n=-\infty \\ n \neq 0}}^{n=\infty} \frac{1}{n} e^{jn \frac{\pi}{T} t}$$

P 15.9-6**Solution:**

$$\begin{aligned}
 F(\omega) &= \int_{-2}^2 e^{-j\omega t} dt - \int_{-1}^1 e^{-j\omega t} dt = \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{-2}^2 - \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{-1}^1 = \frac{1}{j\omega} (e^{j2\omega} - e^{-j2\omega}) - \frac{1}{j\omega} (e^{j\omega} - e^{-j\omega}) \\
 &= \frac{2}{\omega} (\sin 2\omega - \sin \omega)
 \end{aligned}$$

P 16.3-4**Solution:**

Use Table 16-3.2 to obtain the transfer function of a fourth-order Butterworth high-pass filter having a cutoff frequency equal to 1 rad/s and a dc gain equal to 5.

$$H_n(s) = \frac{5s^4}{(s^2 + 0.765s + 1)(s^2 + 1.848s + 1)}$$

Frequency scaling can be used to adjust the cutoff frequency 400 hertz = 2513 rad/s:

$$H_H(s) = \frac{5 \cdot \left(\frac{s}{2513}\right)^4}{\left(\left(\frac{s}{2513}\right)^2 + 0.765\left(\frac{s}{2513}\right) + 1\right)\left(\left(\frac{s}{2513}\right)^2 + 1.848\left(\frac{s}{2513}\right) + 1\right)} = \frac{5 \cdot s^4}{(s^2 + 1922.4s + 2513^2)(s^2 + 4644.025s + 2513^2)}$$

P 16.4-1**Solution**

The transfer function is

$$\begin{aligned}
 H(s) = \frac{V_o(s)}{V_s(s)} &= \frac{\frac{sL \times \frac{1}{Cs}}{sL + \frac{1}{Cs}}}{\frac{sL \times \frac{1}{Cs}}{sL + \frac{1}{Cs}} + R} = \frac{\frac{sL}{s^2 LC + 1}}{\frac{sL}{s^2 LC + 1} + R} = \frac{sL}{s^2 LCR + sL + R} = \frac{\frac{s}{RC}}{s^2 + \frac{s}{RC} + \frac{1}{LC}}
 \end{aligned}$$

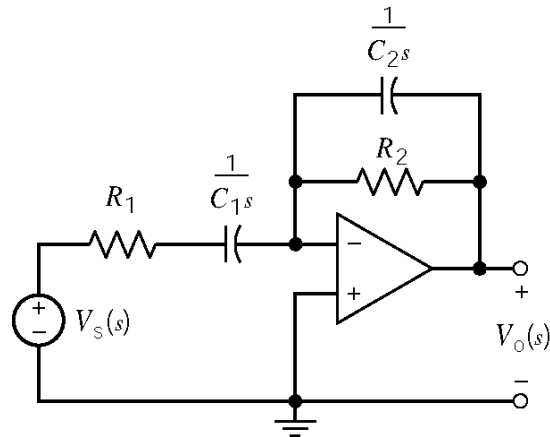
so

$$K = 1, \omega_0^2 = \frac{1}{LC} \quad \text{and} \quad \frac{1}{RC} = \frac{\omega_0}{Q} \Rightarrow Q = RC\omega_0 = R\sqrt{\frac{C}{L}}$$

Pick $C = 1 \mu\text{F}$. Then $L = \frac{1}{C\omega_0^2} = 0.4 \text{ H}$ and $R = Q\sqrt{\frac{L}{C}} = 632 \Omega$

P 16.4-11

Solution:



$$H(s) = \frac{V_o(s)}{V_s(s)} = - \frac{R_2 \parallel \frac{1}{C_2 s}}{R_1 + \frac{1}{C_1 s}} = - \frac{\frac{R_2}{C_2 s + 1}}{\frac{R_1 C_1 s + 1}{C_1 s}} = \frac{-\frac{1}{R_1 C_2} s}{s^2 + \left(\frac{1}{R_1 C_1} + \frac{1}{R_2 C_2} \right) s + \frac{1}{R_1 R_2 C_1 C_2}}$$

Substituting the element values $R_1 = 200 \Omega$, $R_2 = 200 \text{ k}\Omega$, $C_1 = 0.4 \mu\text{F}$ and $C_2 = 50 \text{ pF}$ we determine the center frequency and bandwidth of the filter to be

$$\omega_0 = \frac{1}{\sqrt{R_1 R_2 C_1 C_2}} = 35.4 \text{ k rad/sec} = 2\pi(5.63 \text{ kHz})$$

$$BW = \frac{\omega_0}{Q} = \frac{1}{R_1 C_1} + \frac{1}{R_2 C_2} = 112.5 \text{ k rad/s} = 2\pi(17.9 \text{ kHz})$$